

連続時間系フィルタの自動調整

参照フィルタを用いた周波数特性の自動調整

素子値の一樣偏差の影響

$$T_0(s)=F[sC_i, G_j]$$

一樣偏差: $C_i \rightarrow k_C C_i \quad G_j \rightarrow k_G G_j$

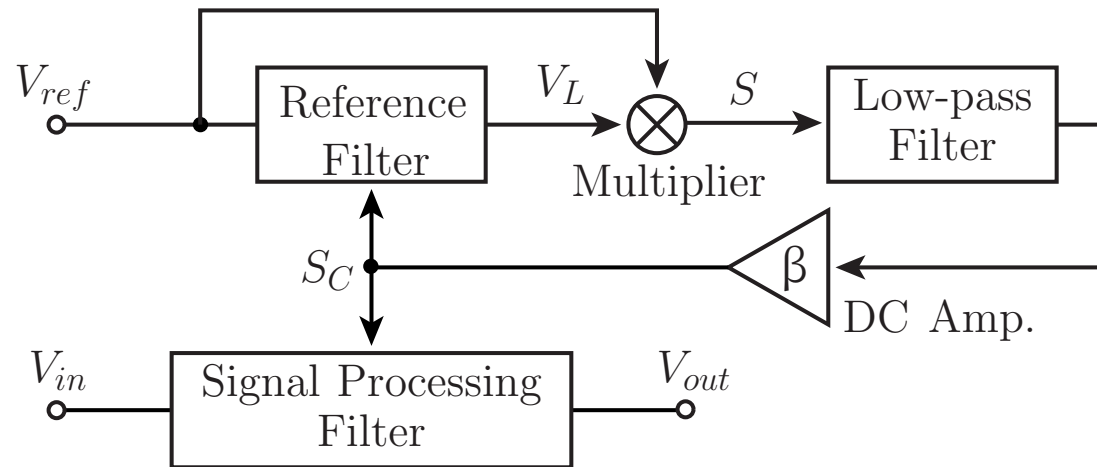
$$T(s)=F[s k_C C_i, k_G G_j]$$

インピーダンス・スケーリング

$$T(s)=F\left[s \frac{k_C}{k_G} C_i, G_j\right]=T_0\left(s \frac{k_C}{k_G}\right)$$

周波数スケーリング

周波数偏差自動調整回路



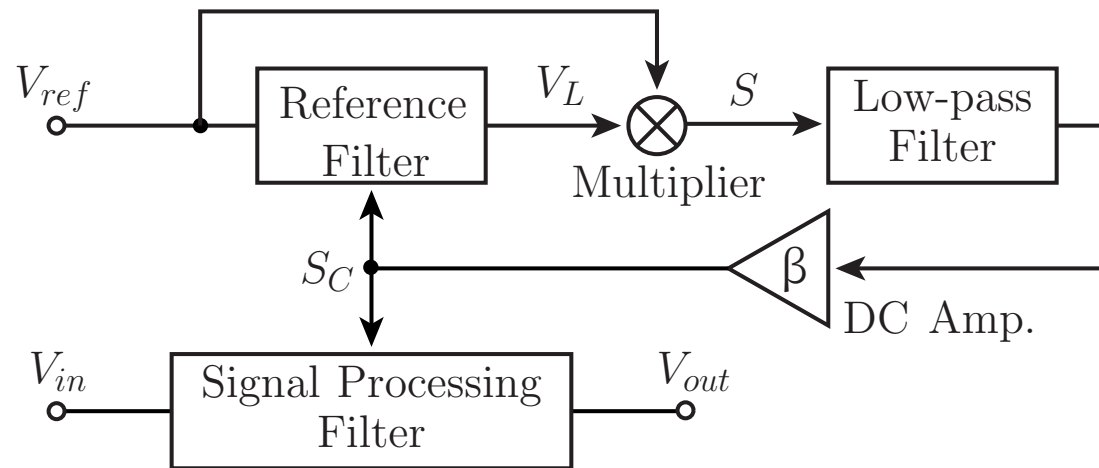
$$V_{ref} = V_m \sin \omega_{ref} t$$

$$V_L = A V_m \sin(\omega_{ref} t + \theta)$$

$$S = M(V_{ref} \times V_L) = M \{ V_m \sin \omega_{ref} t \times A V_m \sin(\omega_{ref} t + \theta) \}$$

$$= \frac{1}{2} M A V_m^2 \{ \cos \theta - \cos(2\omega_{ref} t + \theta) \}$$

$$S_C = \frac{1}{2} \beta M A V_m^2 \cos \theta$$



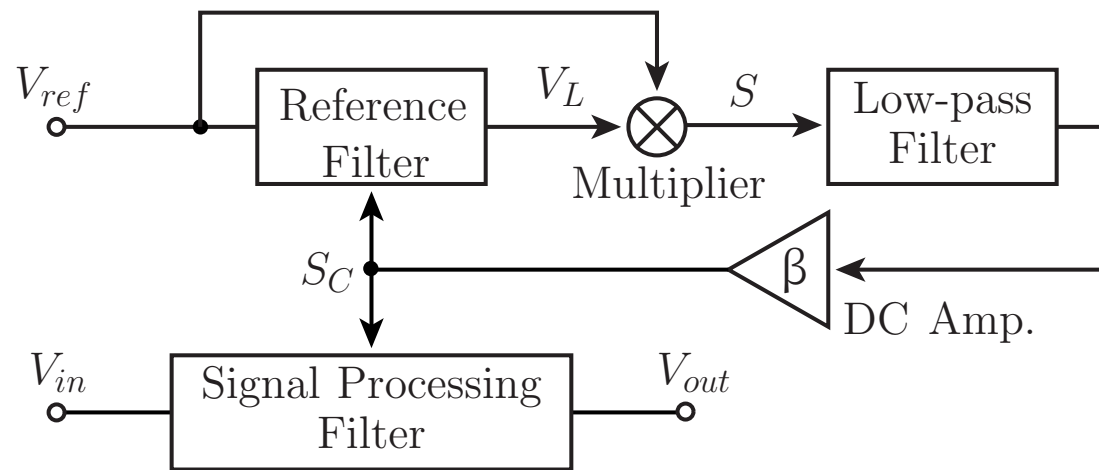
$$S_C = \frac{1}{2} \beta M A V_m^2 \cos \theta$$

$$\beta \rightarrow \infty$$

:有限値に確定

$$\cos \theta = \frac{2 S_C}{\beta M A V_m^2}$$

$$\cos \theta \rightarrow 0$$



参照用フィルタの伝達関数:

$$T_{L2}(s) = \frac{K\omega_0^2}{s^2 + \frac{\omega_0}{Q}s + \omega_0^2}$$

$$\theta = -\tan^{-1} \frac{\omega_{ref}\omega_0}{Q(\omega_0^2 - \omega_{ref}^2)}$$

$$\theta = -90^\circ \rightarrow \omega_0 = \omega_{ref}$$

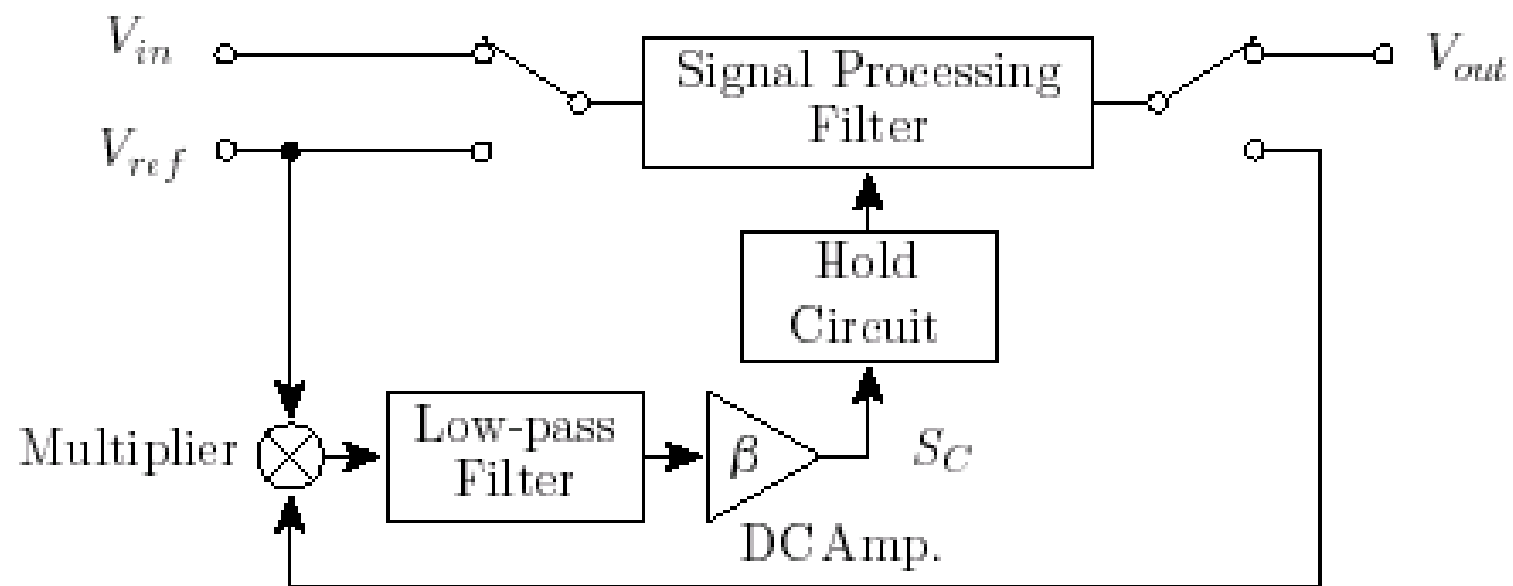
参照用フィルタの素子偏差の割合

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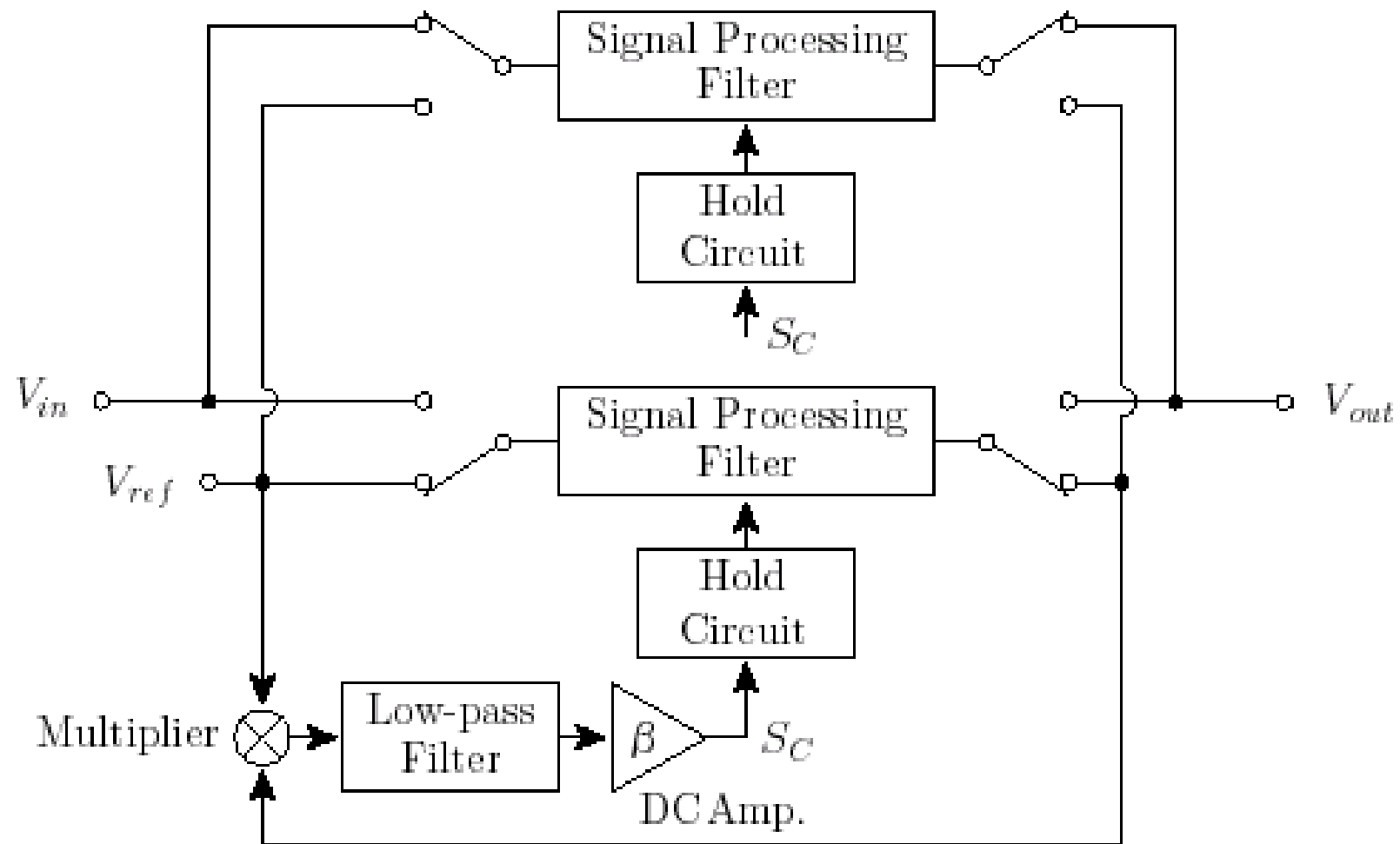
信号処理用フィルタの素子偏差の割合

参照フィルタを用いない周波数特性の自動調整

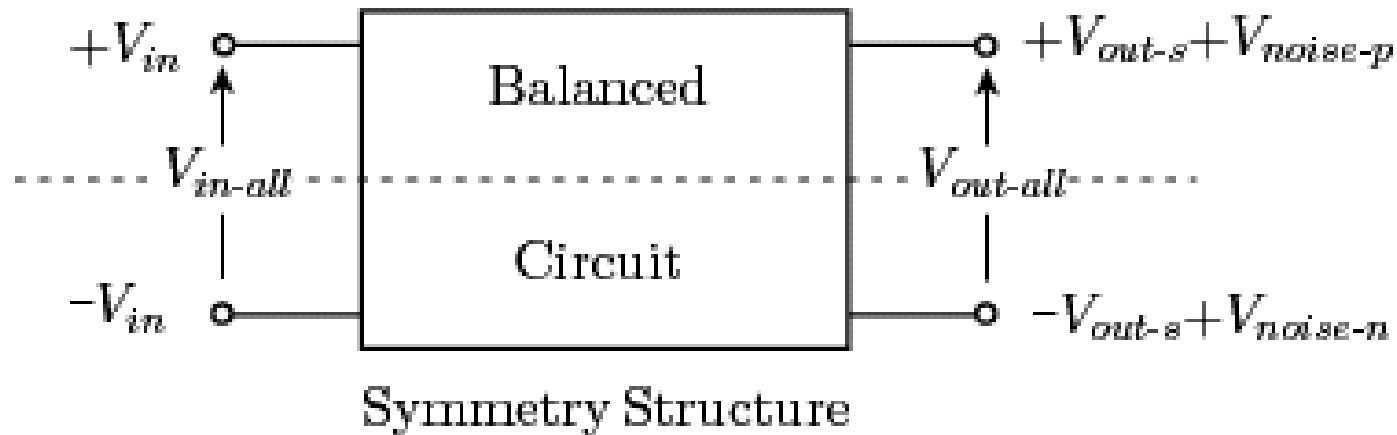
間欠型自動調整



非間欠型自動調整



Balanced-type structure



$$V_{out-p} = +V_{out-s} + V_{noise-p}$$

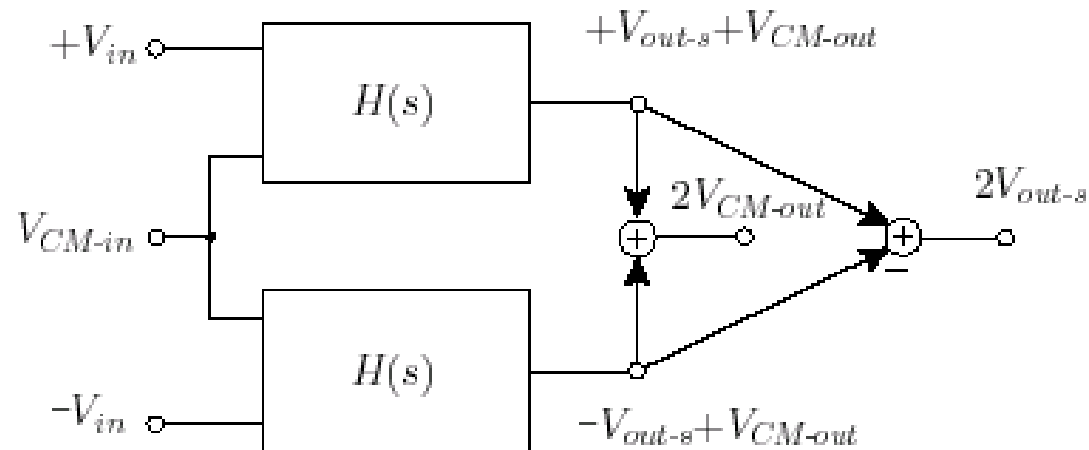
$$V_{out-n} = -V_{out-s} + V_{noise-n}$$

$$V_{noise-p} = +V_{noise-d} + V_{noise-c}$$

$$V_{noise-n} = -V_{noise-d} + V_{noise-c}$$

$$V_{out-all} = V_{out-p} - V_{out-n} = 2(V_{out-s} + V_{noise-d})$$

直接型自動調整



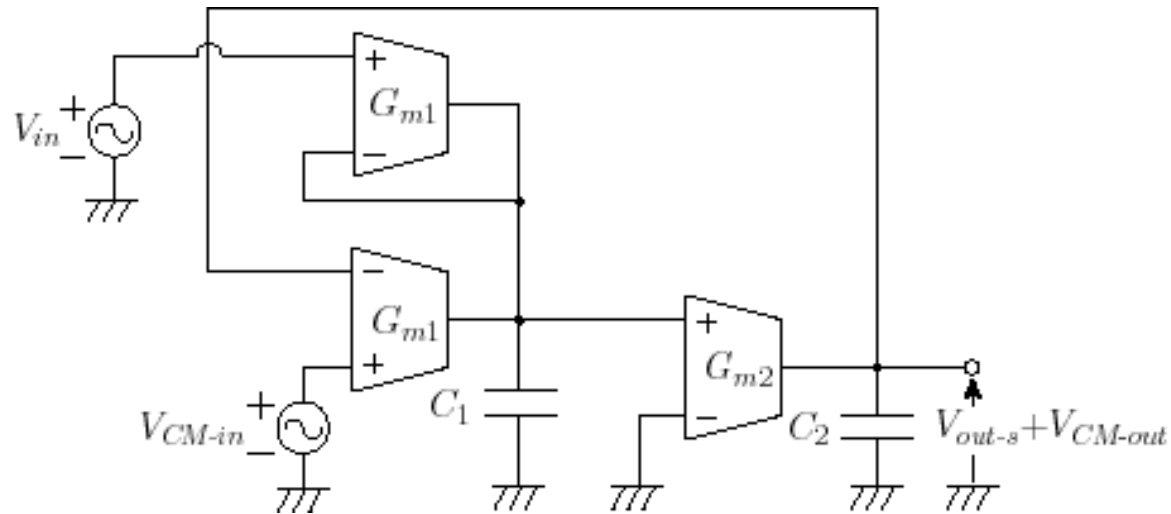
$$V_{out-p} = +V_{out-s} + V_{CM-out} = H_1(s)V_{in} + H_2(s)V_{CM-in}$$

$$V_{out-n} = -V_{out-s} + V_{CM-out} = -H_1(s)V_{in} + H_2(s)V_{CM-in}$$

$$V_{out-p} - V_{out-n} = 2H_1(s)V_{in}$$

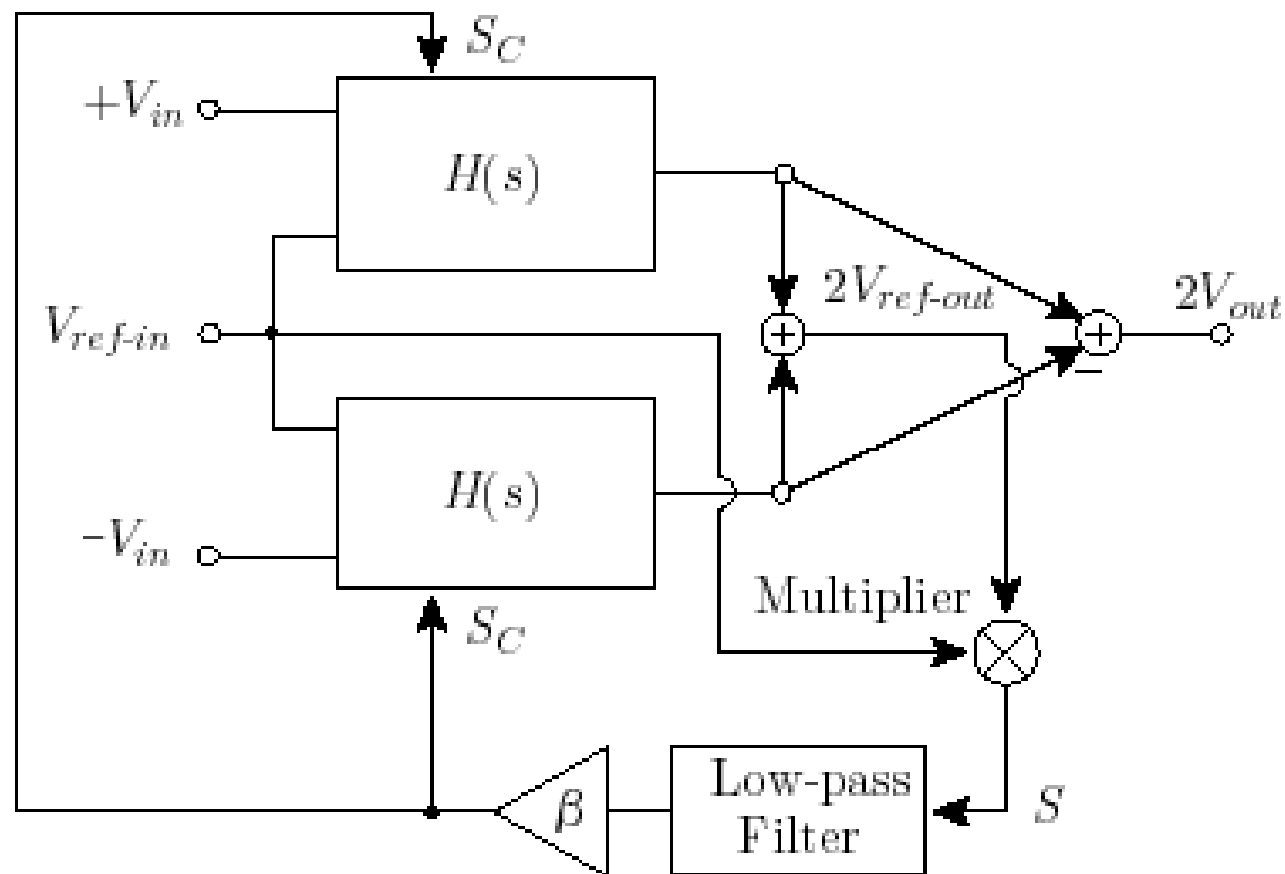
$$V_{out-p} + V_{out-n} = 2H_2(s)V_{CM-in}$$

フィルタの構造

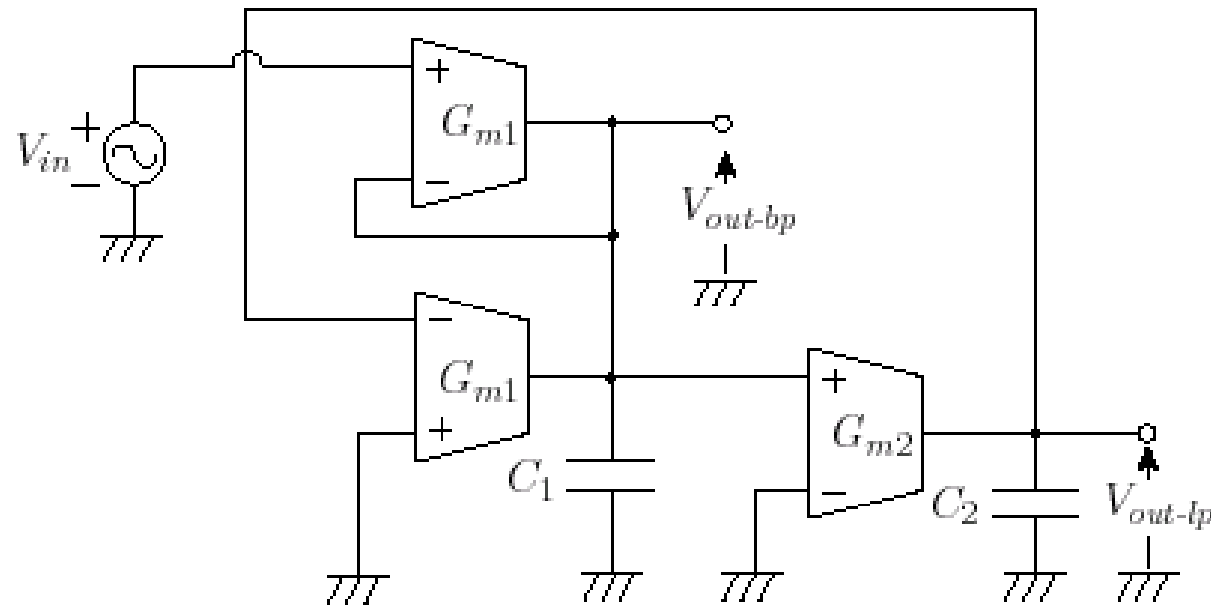


$$H_1(s)=H_2(s)=\frac{\frac{G_{m1}G_{m2}}{C_1C_2}}{s^2+\frac{G_{m1}}{C_1}s+\frac{G_{m1}G_{m2}}{C_1C_2}}$$

直接型自動調整の原理



フィルタのクオリティファクタの自動調整



$$V_{out-bp} = \frac{\frac{G_{m1}s}{C_1}}{s^2 + \frac{G_{m1}}{C_1}s + \frac{G_{m1}G_{m2}}{C_1C_2}} V_{in} = \frac{F_1(s)}{1 + F_1(s) + F_1(s)F_2(s)} V_{in}$$

$$\text{積分回路の特性 : } F_i(s) = a_i \frac{\omega_t}{s} e^{s\tau(\omega)}$$

$$V_{out-bp} = \frac{F_1(s)}{1 + F_1(s) + F_1(s)F_2(s)} V_{in}$$

$$F_i(s) = a_i \frac{\omega_t}{s} e^{s\tau(\omega)}$$

$$T_{bp}(s) = \frac{H_0 \frac{\omega_0}{Q} s}{s^2 + \frac{\omega_0}{Q} s + \omega_0^2}$$

$$\omega_a \approx \omega_0$$

$$Q_a \approx \frac{Q}{1 - 2Q\omega_0\tau}$$

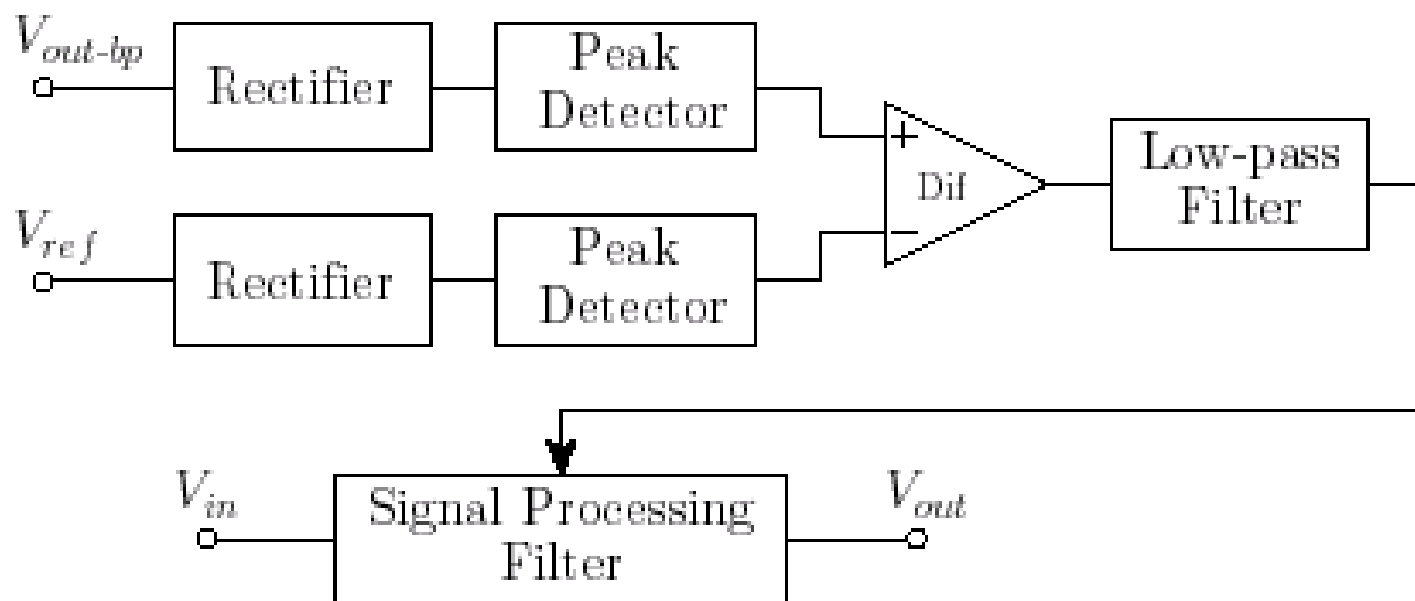
$$H_a \approx \frac{H_0}{1 - 2Q\omega_0\tau}$$

$$Q \approx \frac{Q_a}{2Q_a\omega_0\tau + 1} \Rightarrow H_a \approx H_0(2Q_a\omega_0\tau + 1)$$

$$\tau(\omega) \approx \tau(\omega_0)$$

クオリティファクタの自動調整の原理

ω_0 : V_{ref} の角周波数



例題

参照用低域通過フィルタ

$$T_{L2}(s) = \frac{K\omega_0^2}{s^2 + \frac{\omega_0}{Q}s + \omega_0^2} \quad K=1, \quad \omega_0=2\pi\times 1000\text{rad/s}, \quad Q=5$$

乗算回路

$$M=1\text{V}^{-1}$$

直流増幅回路

$$\beta=100$$

参照入力信号振幅

$$V_m=0.2\text{V}$$

$$\cos\theta = \frac{2S_C}{\beta M A V_m^2}$$

$$A=?$$

$$S_C=?$$

$\omega_{ref} \approx \omega_0$ とすれば

$$A = |T_{L2}(j\omega_{ref})| = \left| \frac{K\omega_0^2}{-\omega_{ref}^2 + j\frac{\omega_0}{Q}\omega_{ref} + \omega_0^2} \right| \approx Q$$

$g_m = \alpha S_C$ かつ $g_m \propto \omega_0$ とすれば

$$\omega_{00} \rightarrow S_{C0}$$

$$\omega_0 \rightarrow S_C$$

$$\frac{\omega_0}{\omega_{00}} = \frac{S_C}{S_{C0}}$$

$$S_C = \frac{\omega_0}{\omega_{00}} S_{C0}$$

$$M=1\text{V}^{-1} \quad \beta=100 \quad V_m=0.2\text{V} \quad A \approx Q=5$$

$$\omega_{00}=2\pi \times 800 \text{rad/s}, \quad S_{C0}=0.1\text{V} \rightarrow S_C=0.125\text{V}$$

$$\cos \theta = \frac{2S_C}{\beta M A V_m^2} = 0.03125 \quad \theta = -88.2^\circ$$

$$\theta = -\tan^{-1} \frac{\omega_{ref} \omega_0}{Q(\omega_0^2 - \omega_{ref}^2)}$$

$$\omega_0 = \frac{-1 + \sqrt{1 + 4Q^2 \tan^2 \theta}}{2Q \tan \theta} \omega_{ref} = 0.992 \omega_{ref}$$