## Advanced Signal Processing – Final Exam

Prof. A. Nishihara

1. Consider a digital filter shown in Figure 1.

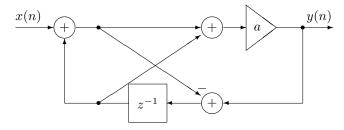


Figure 1: A filter

- (a) Derive its transfer function.
- (b) What is the range of the parameter a for the filter to be stable?
- (c) What are the amplitude responses at  $\omega = 0$  and at  $\omega = \pi$ ?
- (d) What is the filter type (lowpass, highpass,  $\cdots$ )?
- (e) Sketch the rough frequency-amplitude response.
- (f) What happens to the transfer function when the direction of subtraction is reversed.
- (g) What are the critical path and iteration bound? Use  $T_M$  for multiplication time and  $T_A$  for additon/subtraction time.
- 2. Let x(n) be periodic with period N (N is the smallest interger such that x(n) = x(n+N)). Let y(n) be the M-fold decimated version (y(n) = x(Mn)). Show that y(n) is periodic, i.e. there exists  $L < \infty$  such that y(n) = y(n+L) for all n. Assuming no further knowledge about the input, what is the smallest value of L in terms of M and N?
- 3. Consider a sinusoid in white noise

 $x(n) = \sqrt{2}\sin\left(n\omega\right) + e(n).$ 

- (a) Show that the autocorrelation function is  $r(\ell) = \cos(\ell\omega) + \sigma_e^2 \delta(\ell)$ .
- (b) Show that the eigenvalues of the  $3 \times 3$  autocorrelation matrix

$$R = \left(\begin{array}{ccc} r(0) & r(1) & r(2) \\ r(1) & r(0) & r(1) \\ r(2) & r(1) & r(0) \end{array}\right)$$

are

$$\lambda_1 = \sigma_e^2,$$
  

$$\lambda_2 = \sigma_e^2 + 1 - \cos 2\omega,$$
  

$$\lambda_3 = \sigma_e^2 + 2 + \cos 2\omega.$$

- 4. Consider the DFG shown in Figure 2, where the number at each node denotes its execution time.
  - (a) What is the critical path length?
  - (b) What is the iteration bound?
  - (c) Manually retime this DFG to minimize the sampling period.

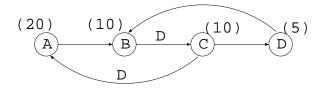


Figure 2: A data flow graph

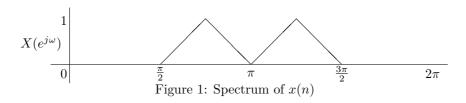
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1. Consider a digital filter described by

$$H(z) = \frac{1}{16}(3 + 8z^{-1} + 6z^{-2} - z^{-4}).$$

- (a) What is the DC gain of this filter?
- (b) What is the filter type (low-pass, high-pass, or  $\cdots$ )?
- (c) What is the relation between  $H(e^{j\omega})$  and  $H(e^{j(\pi-\omega)})$
- (d) What is the particular feature of this filter (What is the class of this filter?)
- (e) Show a structure to realize this transfer function, and show its critical path length (You don't need to minimize the critical path length).
- 2. Consider a signal x(n) whose spectrum  $X(e^{j\omega})$  is given by Figure 1. y(n) is made by y(n) = x(2n).
  - (a) Sketch the spectrum of y(n).
  - (b) Show how you can recover x(n) from y(n).



3. Find the first three terms of the autocorrelation function of the auto-regressive signal

x(n) = 0.8x(n-1) - 0.2x(n-2) + e(n),

where e(n) is a unit-power zero-mean white noise.

4. Write the transfer functions of the three filters shown in Figure 2. State the relationship among the three structures.

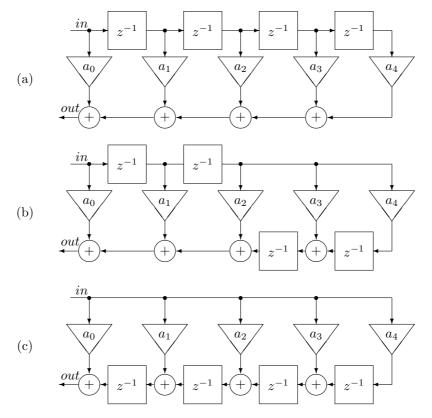


Figure 2: Three Filters

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1. The first row of the 8-point DFT is written as

$$X(0) = x(0) + x(1) + x(2) + x(3) + x(4) + x(5) + x(6) + x(7).$$

- (a) Consider the above as a filter, and write its transfer function.
- (b) Calculate its frequency response.
- (c) What is the filter type (lowpass, highpass,  $\cdots$ )?
- (d) Write the next row of the 8-point DFT.
- (e) Write its transfer function.
- (f) Calculate its frequency response.
- (g) How does the frequency response related to that of the first row?
- 2. For the filter bank shown in Figure 1,
  - (a) Derive the transfer functions  $H_0(z)$  and  $H_1(z)$  of the analysis filter bank.
  - (b) Derive the transfer functions  $F_0(z)$  and  $F_1(z)$  of the synthesis filter bank.
  - (c) Show that the filter bank has perfect reconstruction property.
  - (d) When  $E_0(z) = E_1(z) = 1 + 0.6z^{-1}$ , show transfer functions  $H_0(z), H_1(z), F_0(z)$  and  $F_1(z)$ . Confirm IIR transfer functions are stable. Calculate the overall transfer function  $T(z) = \frac{\hat{X}(z)}{X(z)}$ .
  - (e) What is the main disadvantage of this filter bank?

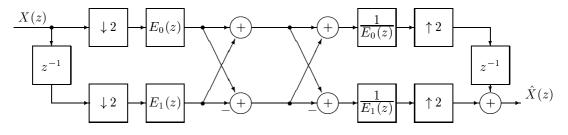


Figure 1: A 2-channel filter bank

3. Consider an FIR fiter

$$y(n) = a_0 x(n) + a_2 x(n-2) + a_3 x(n-3).$$

Assume that the time required for 1 multiply operation is  $T_M$  and the time required for 1 add operation is  $T_A$ .

- (a) Draw a direct-form realization structure of the above filter.
- (b) What is the critical path length?
- (c) Pipeline this filter such that the clock period is  $T_M + T_A$ .
- (d) Draw the broadcast filter architecture for the above filter. What are the advantages and disadvantages of the broadcast architecture compared to the direct-form?

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- 1. For the filter shown in Figure 1,
  - (a) Derive the transfer function H(z) = Y(z)/X(z).
  - (b) What is the filter type (lowpass, highpass,  $\cdots$ )?
  - (c) Show that the transfer function is FIR.
  - (d) Derive the transfer function F(z) = W(z)/X(z).
  - (e) When the input signal is a unit step function u(n), derive w(n) and y(n).
  - (f) What kind of potential problem do you find in the above w(n)?
  - (g) How can you solve that problem by modifying the structure?

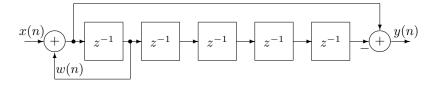


Figure 1: A filter

2. Consider the structure shown in Figure 2, where H(z) is given by

$$H(z) = \frac{(1+z^{-1})^2}{(1-0.5jz^{-1})(1+0.5jz^{-1})}$$

- (a) Show a structure of H(z) using only real coefficients.
- (b) What is the filter type (lowpass, highpass,  $\cdots$ )?
- (c) Show the critical path. Suppose one multiplication time is 200ns and one addition time is 50ns. What is the maximum input sampling frequency?
- (d) Decompose H(z) using two polyphase components and show the structure.
- (e) Use the Noble Identity and show thus modified structure.
- (f) What is the maximum input sampling frequency in this case?

$$\longrightarrow H(z) \longrightarrow \downarrow 2 \longrightarrow$$

Figure 2: A system

3. A measurement gives the signal autocorrelation values r(0) = 1.0, r(1) = 0.6, and r(2) = 0.2. Suppose the signal is generated by a second-order linear prediction filter excited by a white noise. Calculate the two coefficients of the second-order linear predictor and the white noise power.

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- 1. For the filter shown in Fig.1,
  - (a) Derive the transfer function H(z) = Y(z)/X(z).
  - (b) What is the filter type (lowpass, highpass,  $\cdots$ )?
  - (c) When w(n) is made as y(n) x(n), what is the transfer function W(z)/X(z)?
  - (d) Derive the iteration bound using  $T_M$ , one multiplication time, and  $T_A$ , one addition time.

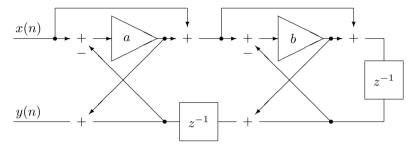


Figure 1: A filter

- 2. Suppose  $H_0(z)$  is a half-band lowpass filter and  $H_1(z)$  is its complementary highpass filter. Consider the structure shown in Fig. 2.
  - (a) Derive the overall transfer function.
  - (b) Draw rough plot of the overall amplitude response. Show the values corresponding to  $a_k$  in the plot.
  - (c) What kind of applications would you find for this circuit?
  - (d) Draw a diagram when the number of channels is further increased.

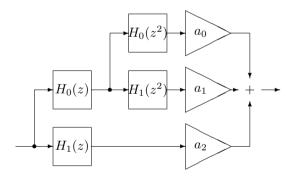


Figure 2: A circuit

3. LMS adaptive filters have coefficient update formula given by

 $H(n+1) = H(n) + \mu e(n)X(n),$ 

where H(n) is the N-tap FIR filter coefficient vector, e(n) is the error, X(n) is the N most recent input signal vector, at time n. The stepsize parameter  $\mu$  is a scalar constant, i.e., uniform coefficient update gain. There is a variable stepsize algorithm to vary  $\mu$  according to the situation. It is further possible to replace this scalar  $\mu(n)$  by a matrix M(n) given by

$$M(n) = \begin{bmatrix} \mu_0(n) & 0 & 0 & \cdots & 0 \\ 0 & \mu_1(n) & 0 & \cdots & 0 \\ 0 & 0 & \mu_2(n) & \cdots & 0 \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & 0 & \cdots & \mu_{N-1}(n) \end{bmatrix}$$

- (a) What will be an advantage of this replacement?
- (b) What kind of strategy can you consider for the choice of  $\mu_k(n)$  to improve the performance of this adaptive filter?