

# 9. Scattering Matrix

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The matrix representations of circuit are addressed focused on the scattering matrix. The topics include:

9.1 Definition of scattering matrix

9.2 Scattering matrix and transmission matrix

9.3 Relation between scattering matrix and impedance matrix

(9.4 Method for determination of scattering matrix)

# Definition

Along a transmission line

$$V(y) = V_i e^{j\beta y} + V_r e^{-j\beta y} = \sqrt{R_c} \left( \frac{V_i}{\sqrt{R_c}} e^{j\beta y} + \frac{V_r}{\sqrt{R_c}} e^{-j\beta y} \right)$$
$$I(y) = \frac{1}{R_c} (V_i e^{j\beta y} - V_r e^{-j\beta y}) = \frac{1}{\sqrt{R_c}} \left( \frac{V_i}{\sqrt{R_c}} e^{j\beta y} - \frac{V_r}{\sqrt{R_c}} e^{-j\beta y} \right) \quad (9.1)$$

Let's define the following quantity.

$$a(y) = \frac{V_i}{\sqrt{R_c}} e^{j\beta y} \rightarrow |a(y)|^2 = \frac{|V_i|^2}{R_c} : \text{incident power}$$

$$b(y) = \frac{V_r}{\sqrt{R_c}} e^{-j\beta y} \rightarrow |b(y)|^2 = \frac{|V_r|^2}{R_c} : \text{reflected power}$$

$$V(y) = \sqrt{R_c} (a(y) + b(y))$$

$$I(y) = \frac{1}{\sqrt{R_c}} (a(y) - b(y)) \quad (9.2)$$

$$a(y) = \frac{V(y) + R_c I(y)}{2\sqrt{R_c}}$$

$$b(y) = \frac{V(y) - R_c I(y)}{2\sqrt{R_c}} \quad (9.3)$$

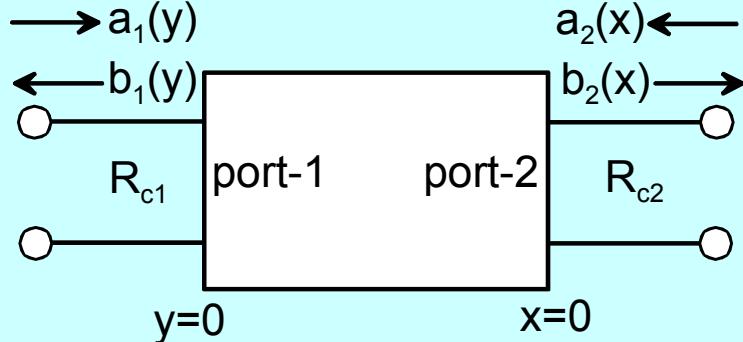
# 2-port circuit

$$a_1(y) = \frac{V_1(y) + R_{c1}I_1(y)}{2\sqrt{R_{c1}}}$$

$$b_1(y) = \frac{V_1(y) - R_{c1}I_1(y)}{2\sqrt{R_{c1}}}$$

$$a_2(x) = \frac{V_2(x) + R_{c2}I_2(x)}{2\sqrt{R_{c2}}}$$

$$b_2(x) = \frac{V_2(x) - R_{c2}I_2(x)}{2\sqrt{R_{c2}}}$$



Definition of scattering matrix:  $a_1=a_1(0)$ , etc

$$\begin{aligned} b_1 &= S_{11}a_1 + S_{12}a_2 \\ b_2 &= S_{21}a_1 + S_{22}a_2 \end{aligned} \quad (9.4)$$

$$\begin{pmatrix} b_1 \\ b_2 \end{pmatrix} = \begin{pmatrix} S_{11} & S_{12} \\ S_{21} & S_{22} \end{pmatrix} \begin{pmatrix} a_1 \\ a_2 \end{pmatrix}$$

$$\mathbf{b} = \mathbf{S}\mathbf{a} \quad (9.5)$$

# Loss-less properties

Input power is given by

$$|a_1|^2 + |a_2|^2 + \dots + |a_n|^2 = (a_1^*, a_2^*, \dots, a_n^*) \begin{pmatrix} a_1 \\ a_2 \\ \vdots \\ a_n \end{pmatrix} = \tilde{\mathbf{a}}\mathbf{a}$$

$\tilde{\mathbf{a}}$  : transpose conjugate of  $\mathbf{a}$

Loss-less : input power = output power

$$\tilde{\mathbf{a}}\mathbf{a} = \tilde{\mathbf{b}}\mathbf{b}$$

$$\mathbf{b} = \mathbf{S}\mathbf{a}$$

$$\tilde{\mathbf{b}} = \tilde{\mathbf{a}}\tilde{\mathbf{S}}$$

$$\tilde{\mathbf{a}}\mathbf{a} - \tilde{\mathbf{b}}\mathbf{b} = \tilde{\mathbf{a}}\mathbf{a} - \tilde{\mathbf{a}}\tilde{\mathbf{S}}\mathbf{S}\mathbf{a} = \tilde{\mathbf{a}}(\mathbf{I} - \tilde{\mathbf{S}}\mathbf{S})\mathbf{a} = 0$$

$$\therefore \tilde{\mathbf{S}}\mathbf{S} = \mathbf{S}\tilde{\mathbf{S}} = \mathbf{I} \quad (9.6)$$

Reciprocity:

$$\frac{b_2}{a_1} \Big|_{a_2=0} = \frac{b_1}{a_2} \Big|_{a_1=0}$$



$$S_{ij} = S_{ji}$$

$$\therefore S_{21} = S_{12} \quad (9.7)$$

# S-matrix and T-matrix

T-matrix

$$\begin{pmatrix} a_1 \\ b_1 \end{pmatrix} = \begin{pmatrix} T_{11} & T_{12} \\ T_{21} & T_{22} \end{pmatrix} \begin{pmatrix} b_2 \\ a_2 \end{pmatrix} \quad (9.8)$$

suitable for cascaded connection

$$\begin{pmatrix} a_1 \\ b_1 \end{pmatrix} = \mathbf{T}_1 \begin{pmatrix} b_2' \\ a_2' \end{pmatrix}$$
$$\begin{pmatrix} b_2' \\ a_2' \end{pmatrix} = \mathbf{T}_2 \begin{pmatrix} b_2 \\ a_2 \end{pmatrix} \quad \therefore \begin{pmatrix} a_1 \\ b_1 \end{pmatrix} = \mathbf{T}_1 \mathbf{T}_2 \begin{pmatrix} b_2 \\ a_2 \end{pmatrix} \quad (9.9)$$

From the definition of S-matrix

$$a_1 = \frac{1}{S_{21}} b_2 - \frac{S_{22}}{S_{21}} a_2$$
$$b_1 = \frac{S_{11}}{S_{21}} b_2 + \left( S_{12} - \frac{S_{11}S_{22}}{S_{21}} \right) a_2$$

$$T_{11} = \frac{1}{S_{21}}, \quad T_{12} = -\frac{S_{22}}{S_{21}}$$
$$T_{21} = \frac{S_{11}}{S_{21}}, \quad T_{22} = S_{12} - \frac{S_{11}S_{22}}{S_{21}} \quad (9.10)$$

# S-matrix and Z-matrix

$$\begin{aligned} a_j &= \frac{V_j + R_{cj} I_j}{2\sqrt{R_{cj}}} & V_j &= \sqrt{R_{cj}}(a_j + b_j) \\ b_j &= \frac{V_j - R_{cj} I_j}{2\sqrt{R_{cj}}} & I_j &= \frac{1}{\sqrt{R_{cj}}}(a_j - b_j) \end{aligned} \quad (9.11)$$

Then,

$$\begin{aligned} \begin{pmatrix} V_1 \\ V_2 \\ \vdots \\ V_n \end{pmatrix} &= \begin{pmatrix} \sqrt{R_{c1}} & & & \\ & \sqrt{R_{c2}} & & \\ & & \ddots & \\ & 0 & & \sqrt{R_{cn}} \end{pmatrix} \begin{pmatrix} a_1 \\ a_2 \\ \vdots \\ a_n \end{pmatrix} + \begin{pmatrix} \sqrt{R_{c1}} & & & \\ & \sqrt{R_{c2}} & & \\ & & \ddots & \\ & 0 & & \sqrt{R_{cn}} \end{pmatrix} \begin{pmatrix} b_1 \\ b_2 \\ \vdots \\ b_n \end{pmatrix} \\ [V] &= [\sqrt{R}]([a] + [b]) \\ [\sqrt{R}] &= \begin{pmatrix} \sqrt{R_{c1}} & & & 0 \\ & \sqrt{R_{c2}} & & \\ & & \ddots & \\ & 0 & & \sqrt{R_{cn}} \end{pmatrix} \end{aligned} \quad (9.12)$$

# S-matrix and Z-matrix

Similarly,

$$[I] = [\sqrt{R}]^{-1}([a] - [b]) \quad (9.13)$$

$$[\sqrt{R}]^{-1} = \begin{pmatrix} \sqrt{R_{c1}} & & 0 \\ & \sqrt{R_{c2}} & \\ & & \ddots \\ 0 & & \sqrt{R_{cn}} \end{pmatrix}^{-1} = \begin{pmatrix} \sqrt{R_{c1}}^{-1} & & 0 \\ & \sqrt{R_{c2}}^{-1} & \\ & & \ddots \\ 0 & & \sqrt{R_{cn}}^{-1} \end{pmatrix}$$

Since  $\boxed{[V]} = [Z]\boxed{[I]}$

$$\boxed{[V]} = [V_i] + [V_r]$$

$$\boxed{[I]} = [R]^{-1}([V_i] - [V_r]) \quad (9.14)$$

$$[V_i] + [V_r] = [Z][R]^{-1}([V_i] - [V_r])$$

$$(1 + [Z][R]^{-1})[V_r] = ([Z][R]^{-1} - 1)[V_i] \quad (9.15)$$

# S-matrix and Z-matrix

$$[V_i] = [\sqrt{R}][a]$$

$$[V_r] = [\sqrt{R}][b]$$

$$[b] = [S][a]$$

Thus,

$$(1 + [Z][R]^{-1})[\sqrt{R}][S] = ([Z][R]^{-1} - 1)[\sqrt{R}]$$

$$\therefore [S] = [\sqrt{R}]^{-1}(1 + [Z][R]^{-1})^{-1}([Z][R]^{-1} - 1)[\sqrt{R}] \quad (9.16)$$

Especially, for  $R_j = R_0$  ( $j=1,2,3\dots n$ )

$$[\sqrt{R}]^{-1} = \frac{1}{\sqrt{R_0}}[1], \quad [R]^{-1} = \frac{1}{R_0}[1], \quad [\sqrt{R}] = \sqrt{R_0}[1]$$

$$[S] = \left(1 + \frac{1}{R_0}[Z]\right)^{-1} \left(\frac{1}{R_0}[Z] - 1\right) = ([Z] + R_0)^{-1}([Z] - R_0) \quad (9.17)$$

# How to find S-matrix elements ?

Simple method:

$$b_1 = S_{11}a_1 \rightarrow S_{11} = \frac{b_1}{a_1} \Big|_{a_2=0}$$
$$b_2 = S_{21}a_1 \rightarrow S_{21} = \frac{b_2}{a_1} \Big|_{a_2=0} \quad (9.18)$$

, but  $a_2=0$  is hard to realize.

Deshamp's method

Terminate port-2 with an impedance  $Z_{L2}$ .

→ Reflection by the impedance  $Z_{L2}$  is given by  $S_2 = \frac{a_2}{b_2} = \frac{Z_{L2} - R_{c2}}{Z_{L2} + R_{c2}}$

$$\frac{b_2}{a_1} = \frac{b_2}{a_2} \frac{a_2}{a_1} = \frac{1}{S_2} \frac{a_2}{a_1} = S_{21} + S_{22} \frac{a_2}{a_1}$$

$$\rightarrow \frac{a_2}{a_1} \left( \frac{1}{S_2} - S_{22} \right) = S_{21}, \quad \frac{a_2}{a_1} = \frac{S_{21}S_2}{1 - S_{22}S_2}$$

$$\text{Reflection at port-1: } S_1 = \frac{b_1}{a_1} = S_{11} + S_{12} \frac{a_2}{a_1} = \frac{S_{11} + (S_{12}S_{21} - S_{11}S_{22})S_2}{1 - S_{22}S_2} \quad (9.20)$$

# How to find S-matrix elements ?

$S_1$  and  $S_2$  : linear transformation

When  $S_2$  moves along a circle  $S_2 = \exp(-2j\beta x)$   
 $S_1$  moves also along another circle.

$$(1 - S_{22}S_2)S_1 = S_{11} + (S_{12}S_{21} - S_{11}S_{22})S_2$$

$$((S_{12}S_{21} - S_{11}S_{22}) + S_{22}S_1)S_2 = S_1 - S_{11}$$

$$S_2 = \frac{S_1 - S_{11}}{S_{22}S_1 + (S_{12}S_{21} - S_{11}S_{22})} \quad (9.21)$$

$|S_2| = 1$  ( $S_2$  is located on a unit circle) gives the following relation.

$$(1 - S_{22}\bar{S}_{22})S_1\bar{S}_1 - ((S_{12}S_{21} - S_{11}S_{22})\bar{S}_{22} + S_{11})\bar{S}_1 - ((S_{12}S_{21} - S_{11}S_{22})S_{22} + \bar{S}_{11})S_1 + (S_{11}\bar{S}_{11} - (S_{12}S_{21} - S_{11}S_{22})(\bar{S}_{12}\bar{S}_{21} - \bar{S}_{11}\bar{S}_{22})) = 0 \quad (9.22)$$

$$\text{center: } \frac{D\bar{S}_{22} + S_{11}}{1 - S_{22}\bar{S}_{22}} = \frac{S_{12}S_{21}\bar{S}_{22}}{1 - S_{22}\bar{S}_{22}} + S_{11} \quad (9.23a)$$

$$\text{radius: } \frac{\sqrt{|D\bar{S}_{22} + S_{11}|^2 - (1 - S_{22}\bar{S}_{22})(S_{11}\bar{S}_{11} - D\bar{D})}}{1 - S_{22}\bar{S}_{22}} = \frac{|S_{12}S_{21}|}{1 - S_{22}\bar{S}_{22}} \quad (9.23b)$$

where  $D = (S_{12}S_{21} - S_{11}S_{22})$

# How to find S-matrix elements ?

By changing the position of movable short, x, we can measure the corresponding reflection  $S_1$ .  
( $S_1$  moves also along another circle.)

e.g.

$$\text{for } S_2 = \pm 1 \quad S_1 = \frac{S_{11} \pm (S_{12}S_{21} - S_{11}S_{22})}{1 \mp S_{22}}$$

By the correspondence between  $S_1$  and  $S_2$ , we can determine  $S_{ij}$ .